



Wave Chaos and Coupling to EM Structures

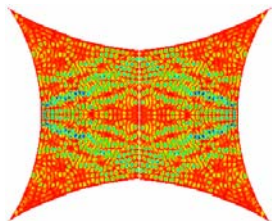


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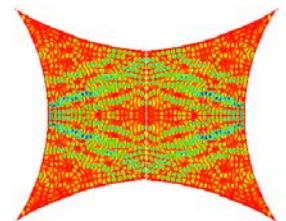
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INSTITUTE FOR RESEARCH IN
ELECTRONICS
& APPLIED PHYSICS



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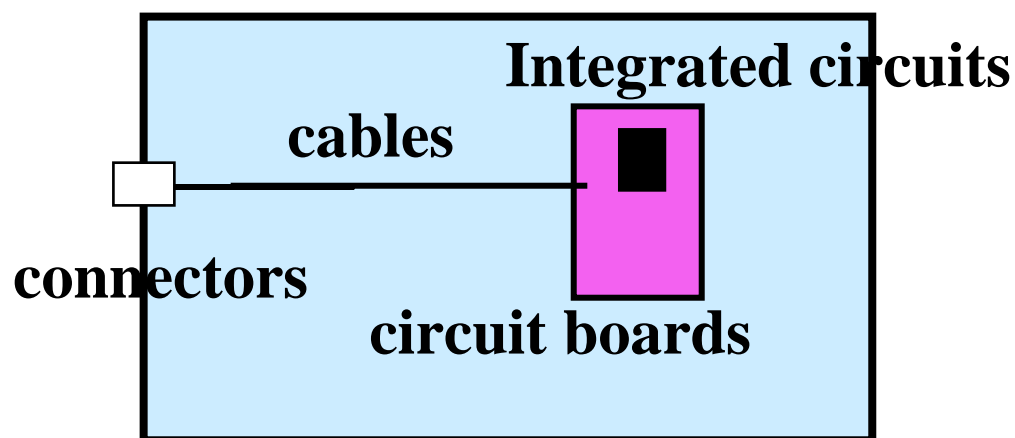


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Electromagnetic Coupling in Computer Circuits

Schematic



- What can be said about coupling without solving in detail the complicated EM problem ?
- Statistical description !

- Coupling of external radiation to computer circuits is a complex process:

apertures
resonant cavities
transmission lines
circuit elements



Outline

- **Part I: Frequency Domain**
 - **Extracting the universal impedance and scattering statistics**
 - **Predictions and tests**
- **Part II: Time Domain**
 - **Model**
 - **Predictions**



Part I:

Frequency Domain

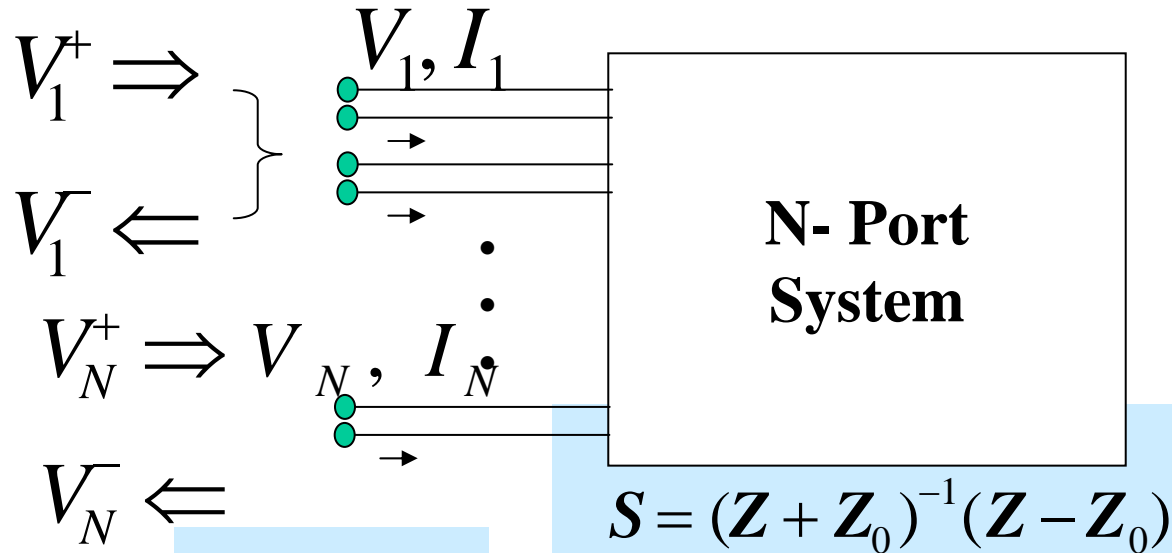


Z and S-Matrices

What is S_{ij} ?

N ports

- voltages and currents,
- incoming and outgoing waves



Z matrix

$$\begin{pmatrix} V_1 \\ V_2 \\ \vdots \\ V_N \end{pmatrix} = \mathbf{Z} \begin{pmatrix} I_1 \\ I_2 \\ \vdots \\ I_N \end{pmatrix}$$

voltage current

S matrix

$$\begin{pmatrix} V_1^- \\ V_2^- \\ \vdots \\ V_N^- \end{pmatrix} = \mathbf{S} \begin{pmatrix} V_1^+ \\ V_2^+ \\ \vdots \\ V_N^+ \end{pmatrix}$$

refl.

inc.

$Z(\omega) S(\omega)$

- Complicated function of frequency
- Details depend very sensitively on unknown parameters.



Statistical Model of $\underline{\underline{Z}}$ Matrix

$$\underline{\underline{Z}}(\omega) = -\frac{j}{\pi} \sum_n \underline{\underline{R}}^{1/2}(\omega_n) \frac{\underline{w}_n \underline{w}_n^T \Delta\omega_n^2}{\omega^2 (1 - jQ^{-1}) - \omega_n^2} \underline{\underline{R}}^{1/2}(\omega_n)$$

$\underline{\underline{Z}}$ = **MxM matrix**

$\underline{\underline{R}}$ = **MxM radiation resistance matrix**

$\Delta\omega_n^2$ = **Mean spectral spacing**

Q = **Quality factor**

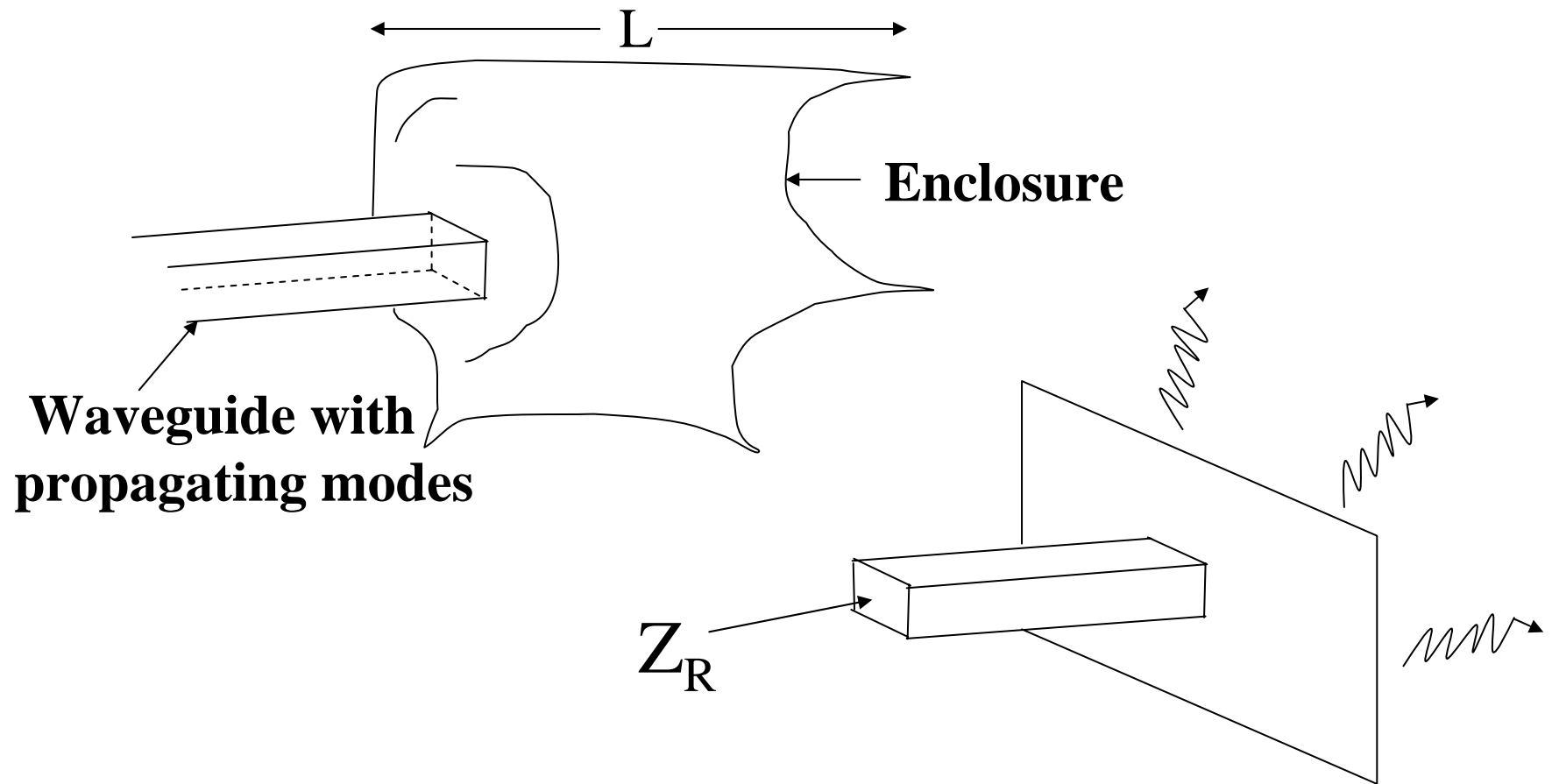
ω_n^2 = **Random Spectrum from RMT**

\underline{w}_n = **M vector of Gaussian random variables**

$$\left\langle \underline{w}_n \underline{w}_n^T \right\rangle = \underline{\underline{1}}$$



Radiation Impedance



$$\underline{\underline{Z}}_R = \underline{\underline{R}}_R + j \underline{\underline{X}}_R$$



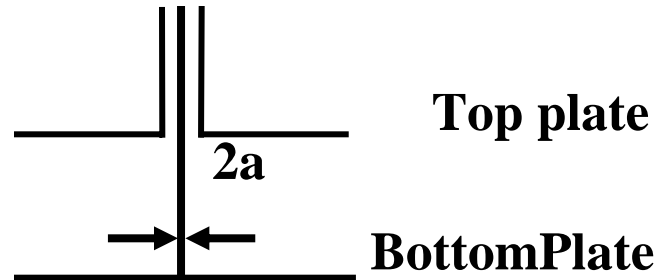
Universal Properties of Impedance

$$\underline{\underline{\xi}} = \underline{\underline{R}}^{-1/2} (\underline{\underline{Z}} - j \underline{\underline{X}}_R) \underline{\underline{R}}^{-1/2}$$

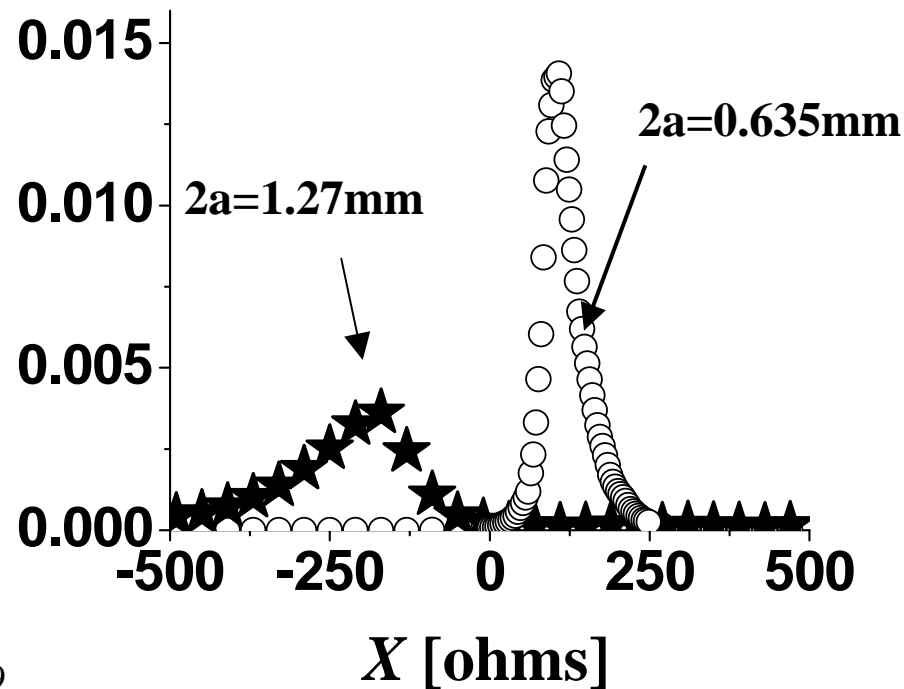
- ξ is universal and obtainable from Random Matrix Theory.
- This applies for (λ/L) small.



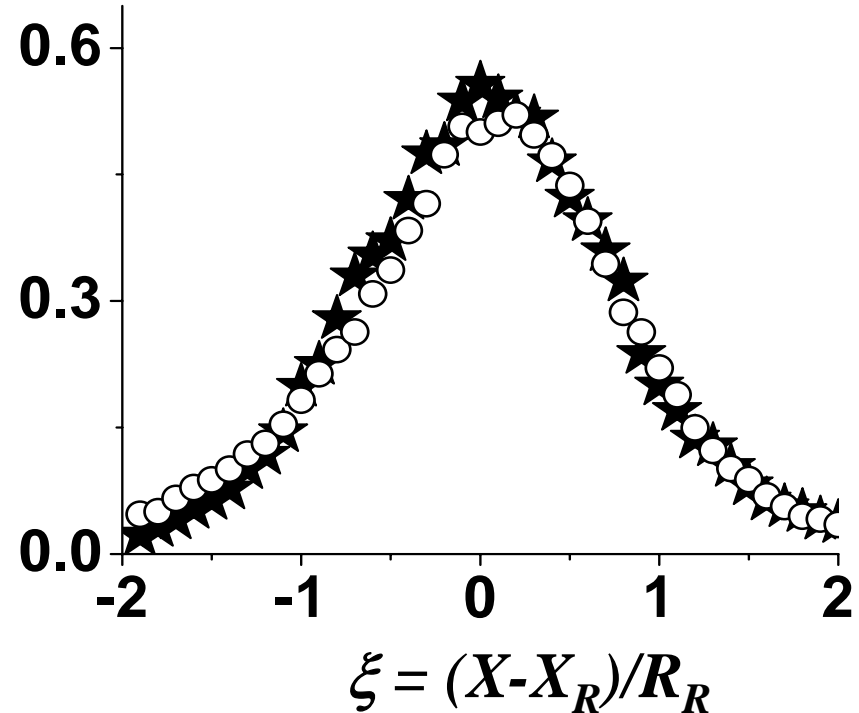
Importance of Normalization



Raw Data



Normalized





Past Results

- **Predictions for statistics of S and Z matrices.**
- **Tests of predictions against numerical solutions of Maxwell's equations.**
- **Tests of predictions against laboratory experiments.**



Predictions Tested

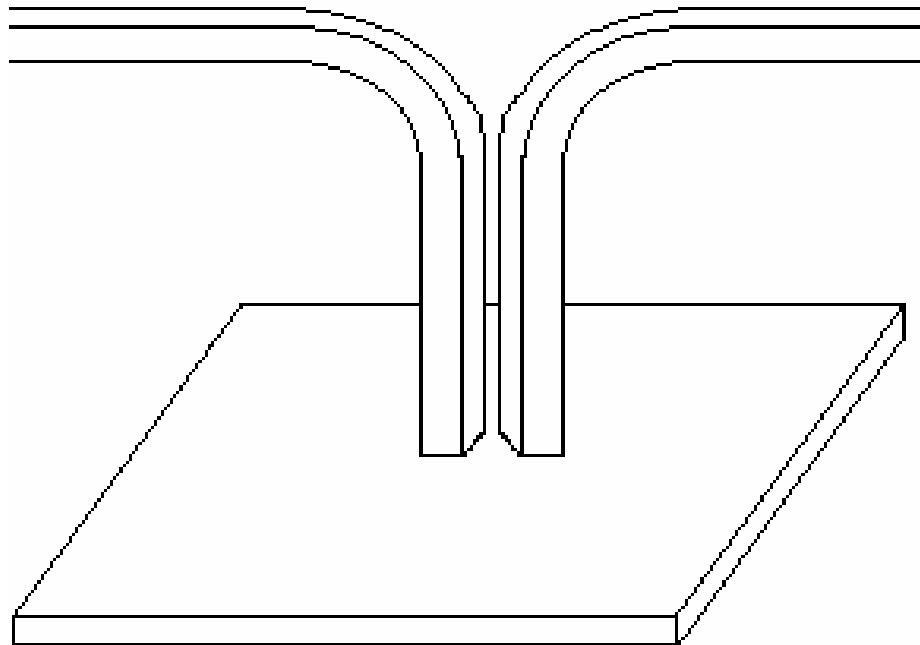
- **M=1**
 - **PDF's of normalized impedances and scattering coefficients as a funct. of loss.**
- **M=2**
 - **Statistics of normalized 2x2 impedance and scattering matrices as a funct. of loss.**
 - **Variance ratios as a funct. of loss [also Fiachetti and Michielsen, Elect. Lett. '03].**

$$VR_Z = \frac{Var[Z_{12}]}{\sqrt{Var[Z_{11}]Var[Z_{22}]}}$$



Some Predictions Not Yet Tested

- Situations where reciprocity does not apply
 - Magnetized ferrite \rightarrow different statistics
- Situations where off-diagonal elements of Z_R are significant.





Some Predictions Not Yet Tested (continued)

- **M>2: E.g.,**

reciprocal

$$\left\langle |S_{ij}|^2 \right\rangle = \begin{cases} \frac{2}{M+1}, & i = j \\ \frac{1}{M+1}, & i \neq j \end{cases}$$

non-reciprocal

$$\left\langle |S_{ij}|^2 \right\rangle = \frac{1}{M}$$



Part II:

Time Domain



Time Domain Model

Frequency Domain

$$Z(\omega) = -\frac{j\omega}{\pi} \sum_n \frac{R_R(\omega_n)}{\omega_n} \frac{\Delta\omega_n^2 w_n^2}{\omega^2(1-jQ^{-1}) - \omega_n^2}$$

w_n - Gaussian Random variables

ω_n - random spectrum

Time Domain

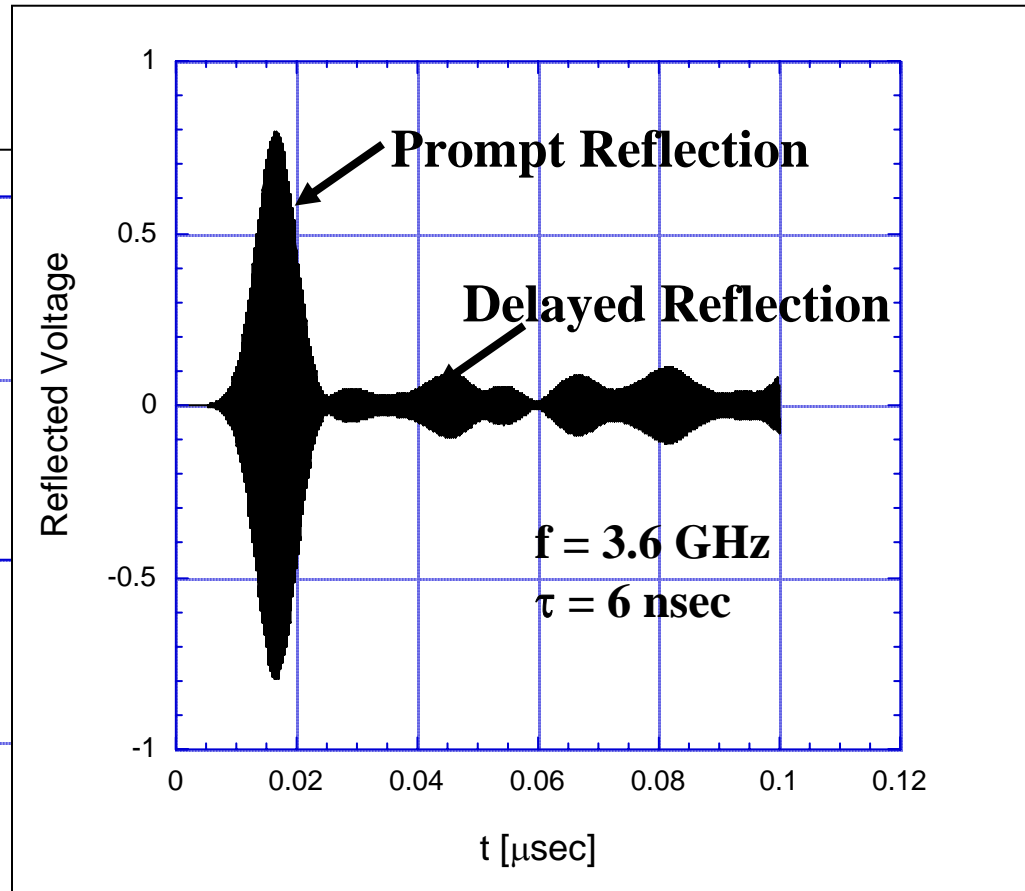
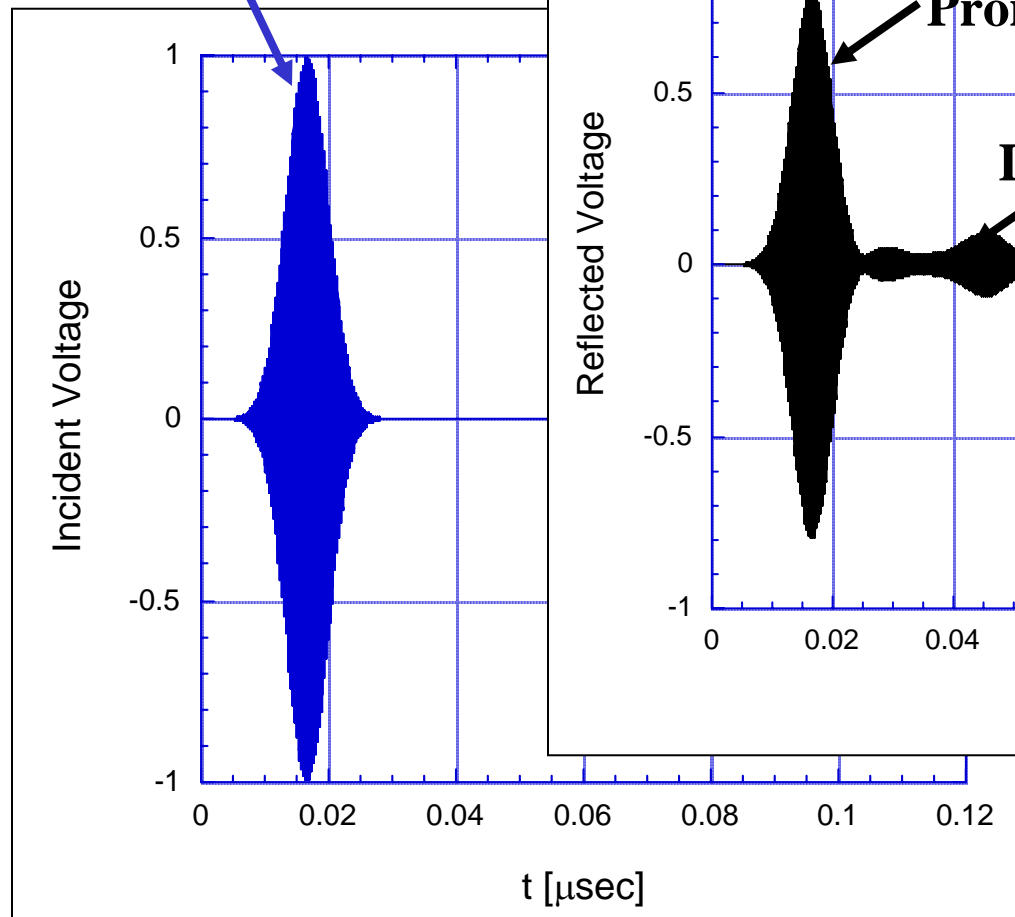
$$\left(\frac{d^2}{dt^2} + 2\nu_n \frac{d}{dt} + \omega_n^2 \right) V_n(t) = -\frac{1}{\pi} \frac{R_R(\omega_n) \Delta\omega_n^2 w_n^2}{\omega_n} \frac{d}{dt} I(t)$$

$$V(t) = \sum_n V_n(t) \quad \nu_n = \frac{\omega_n}{Q}$$



Incident and Reflected Pulses for One Realization

Incident Pulse



Prompt reflection
removed by matching Z_0
to Z_R



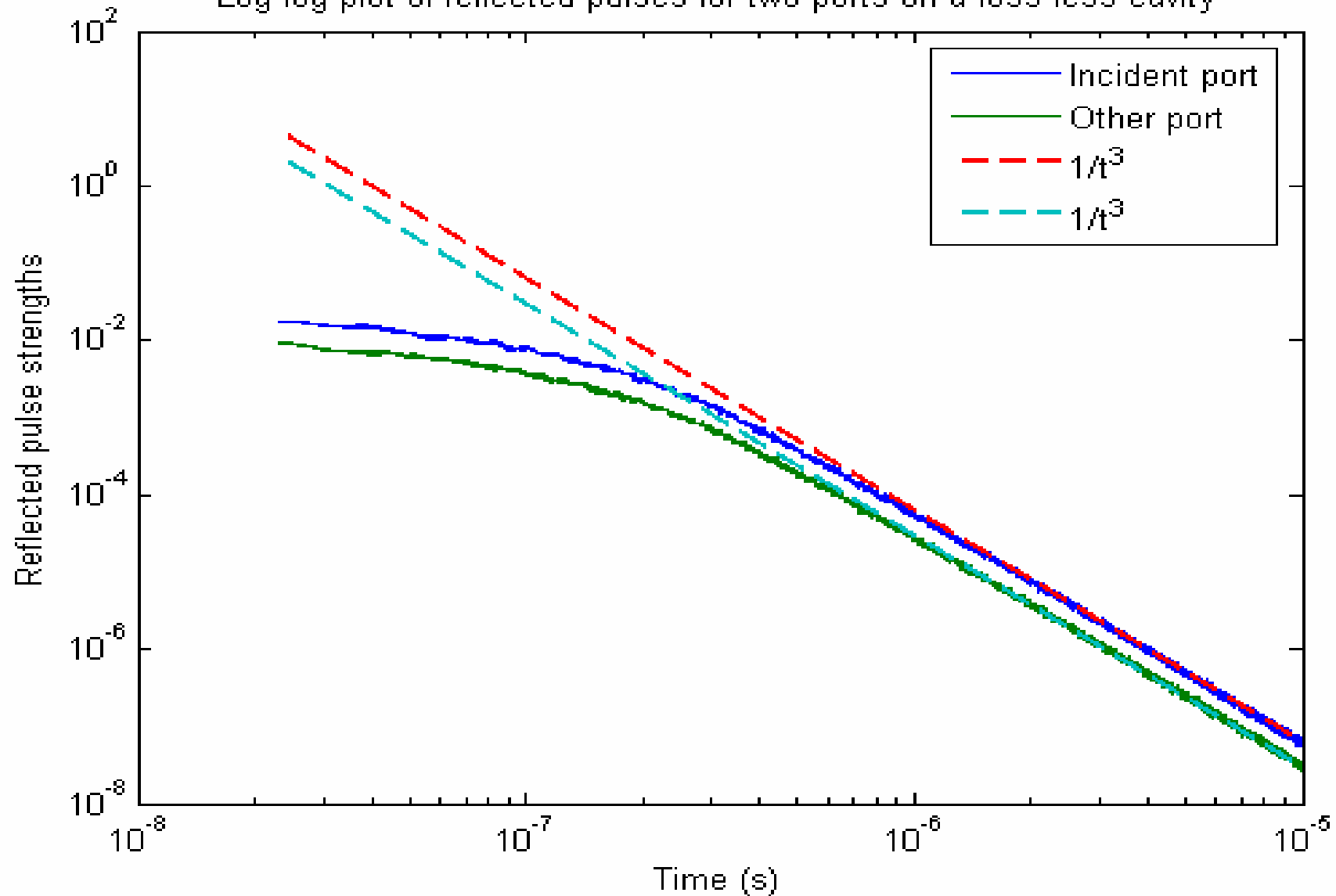
Decay of Port Voltage - Lossless Case

- **One Port with an Incident Pulse:** $\langle V^2(t) \rangle \approx 1 / t^{5/2}$
- **Two Ports Excited Through Port 1,**
 - a) **all ports matched:** $\langle V_1^2(t) \rangle = 2 \langle V_2^2(t) \rangle \approx 1 / t^3$
 - b) **Port 1 matched** $\langle V_1^2(t) \rangle \approx 1 / t^{5/2}$
Port 2 strongly mismatched $\langle V_2^2(t) \rangle \approx 1 / t^{3/2}$
- **N Ports Excited Through Port 1, all ports matched:** $\langle V_1^2(t) \rangle = 2 \langle V_{i \neq 1}^2(t) \rangle \approx 1 / t^{(4+N)/2}$



Simulations of Average Decay

Log-log plot of reflected pulses for two ports on a loss-less cavity



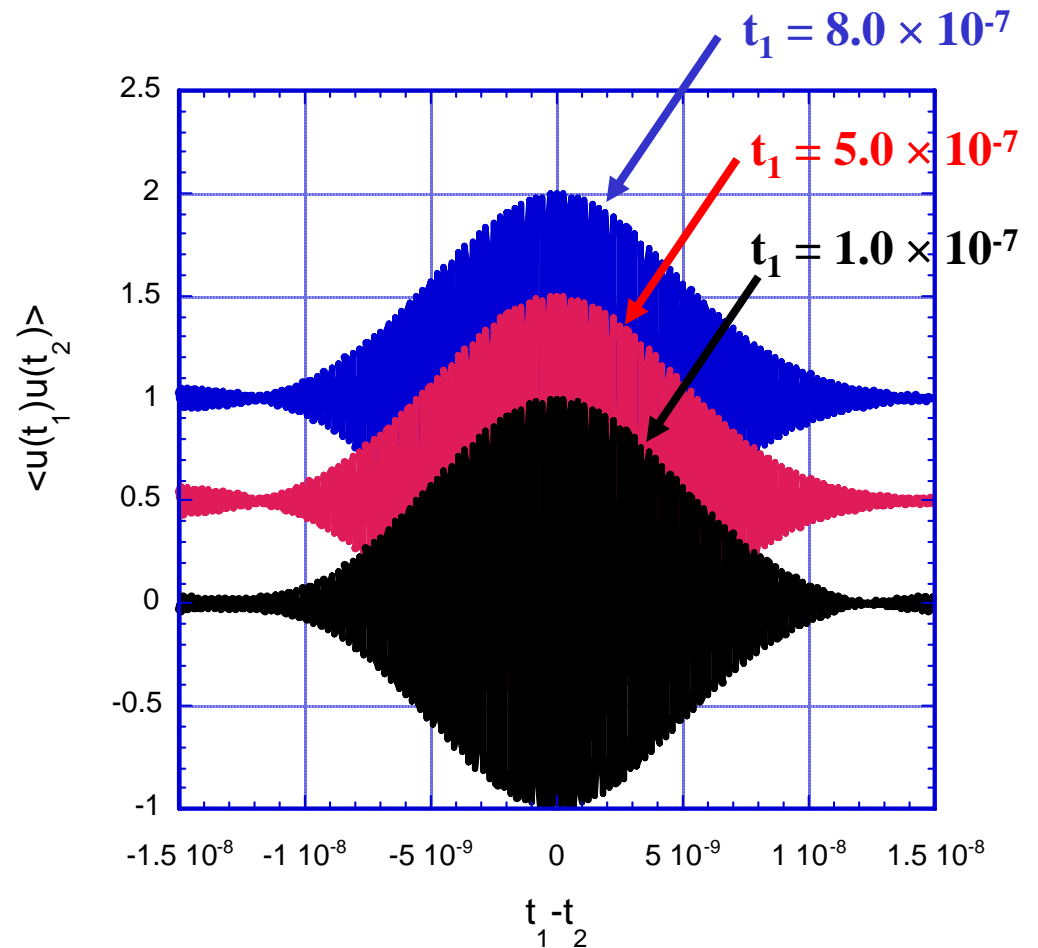
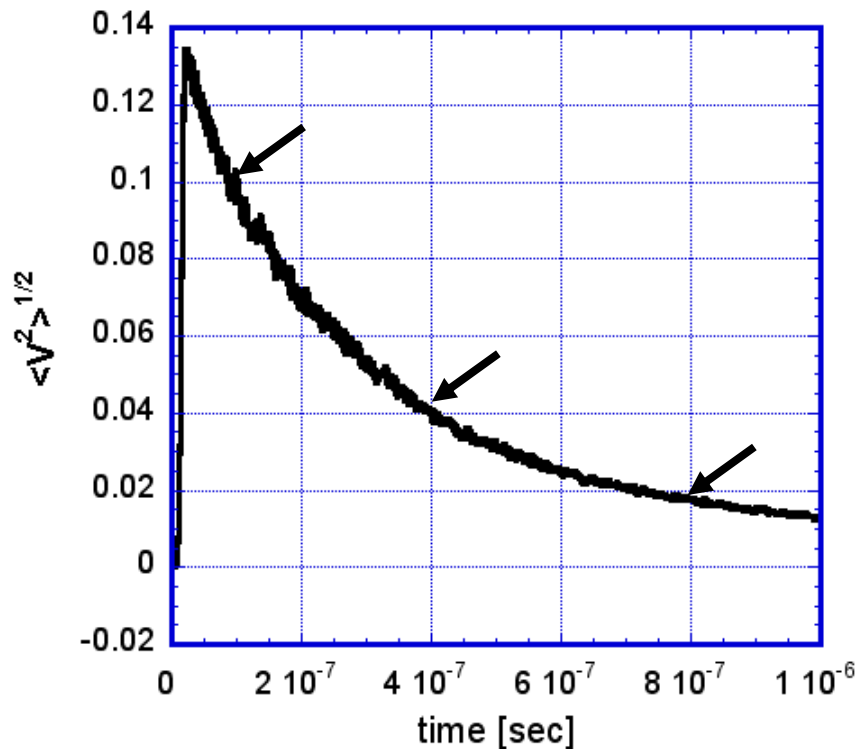


Quasi-Stationary Process

Normalized Voltage

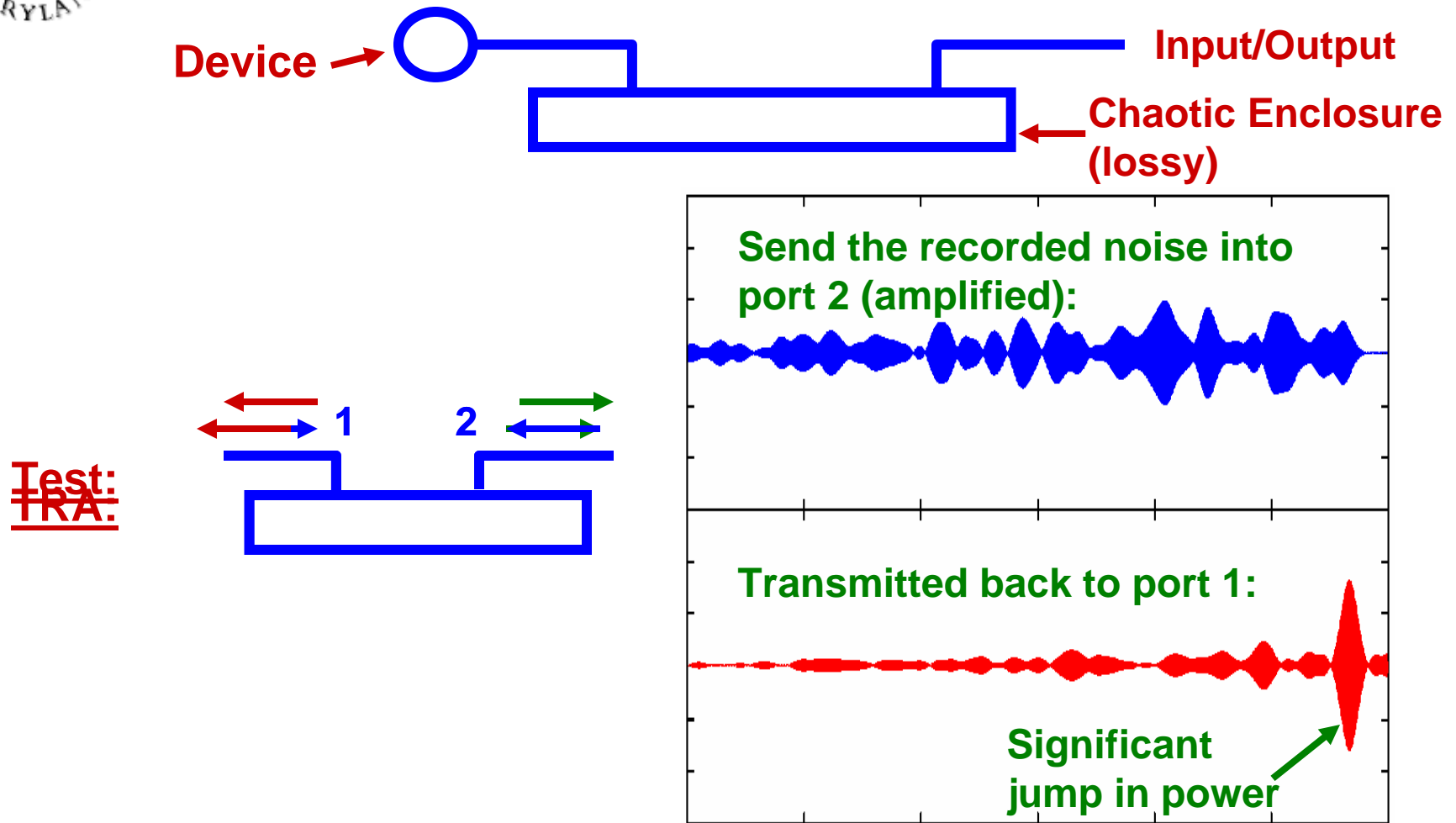
$$u(t) = V(t) / \langle V^2(t) \rangle^{1/2}$$

2-time Correlation Function
(Matches initial pulse shape)





Time Reversal Attack (TRA)



- Issues:**
- ‘Fidelity’ under study
 - Magnetized ferrite: breaks time reversal symmetry



Future Theoretical Work:

- **Corrections for deviations from RMT that occur when $(\lambda/L) \ll 1$ is not well satisfied**
- **Scars - “Anomalous” hot spots**
- **Networks formed by transmission line links**
- **Statistical aspects of coupling of pulsed signals**



Publications

1. ***S. Hemmady, X. Zheng, E. Ott, T. Antonsen, and S. Anlage, Universal Impedance Fluctuations in Wave Chaotic Systems, Phys. Rev. Lett. 94, 014102 (2005).***
2. ***S. Hemmady, X. Zheng, T. Antonsen, E. Ott, and S. Anlage, Universal Statistics of the Scattering Coefficient of Chaotic Microwave Cavities, Phys. Rev. E 71***
3. ***X. Zheng, T. Antonsen, E. Ott, Statistics of Impedance and Scattering Matrices in Chaotic Microwave Cavities: Single Channel Case, Electromagnetics 26, 3 (2006).***
4. ***X. Zheng, T. Antonsen, E. Ott, Statistics of Impedance and Scattering Matrices of Chaotic Microwave Cavities with Multiple Ports, Electromagnetics 26, 37 (2006).***
5. ***X. Zheng, S. Hemmady, T. Antonsen, S. Anlage, and E. Ott, Characterization of Fluctuations of Impedance and Scattering Matrices in Wave Chaotic Scattering, Phys. Rev. E 73 , 046208 (2006).***
6. ***S. Hemmady, X. Zheng, T. Antonsen, E. Ott, S. Anlage, Universal Properties of 2-Port Scattering, Impedance and Admittance Matrices of Wave Chaotic Systems, Phys. Rev. E. submitted.***
7. ***S. Hemmady, X. Zheng, T. Antonsen, E. Ott and S. Anlage, Aspects of the Scattering and Impedance Properties of Chaotic Microwave Cavities, Acta Physica Polonica A 109, 65 (2006).***



Photo by Tom Antonsen

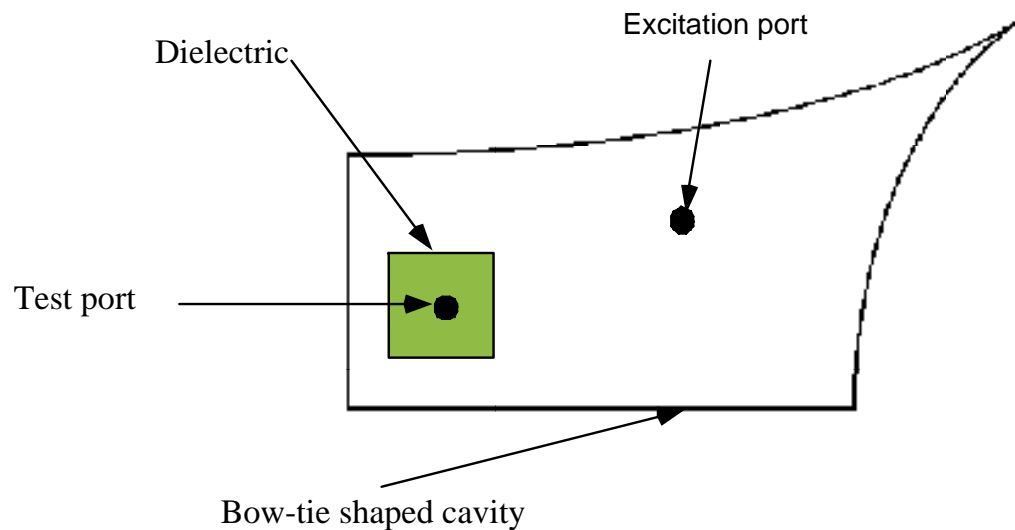


Part III:

Open Problems



More Complexity in the Scatterer



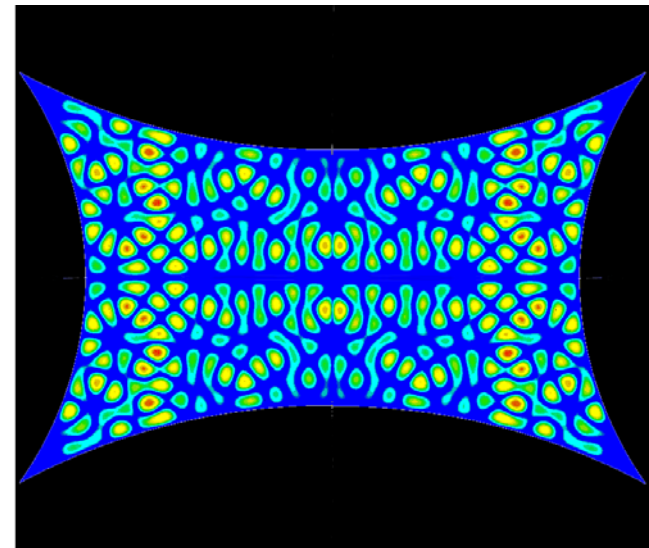
- Can be addressed
 - theoretically
 - numerically
 - experimentally

Features:

Ray splitting

Losses

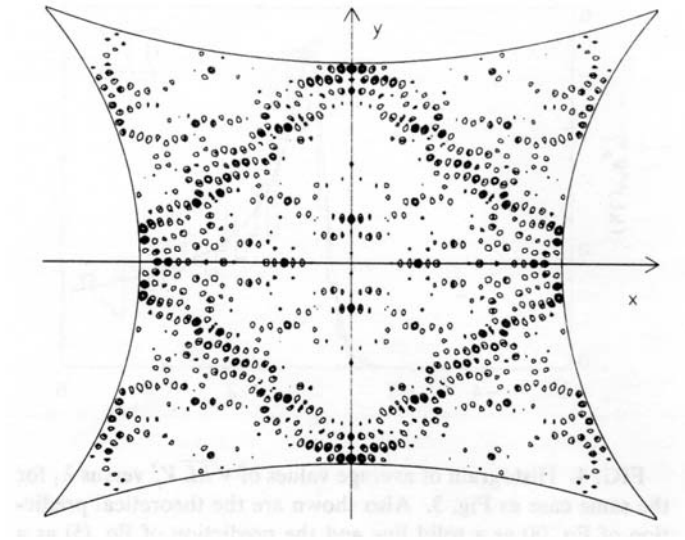
Additional complications can be added





Role of Scars?

- **Eigenfunctions that do not satisfy the random plane wave assumption**
- **Scars are not treated by either random matrix or chaotic eigenfunction theory**



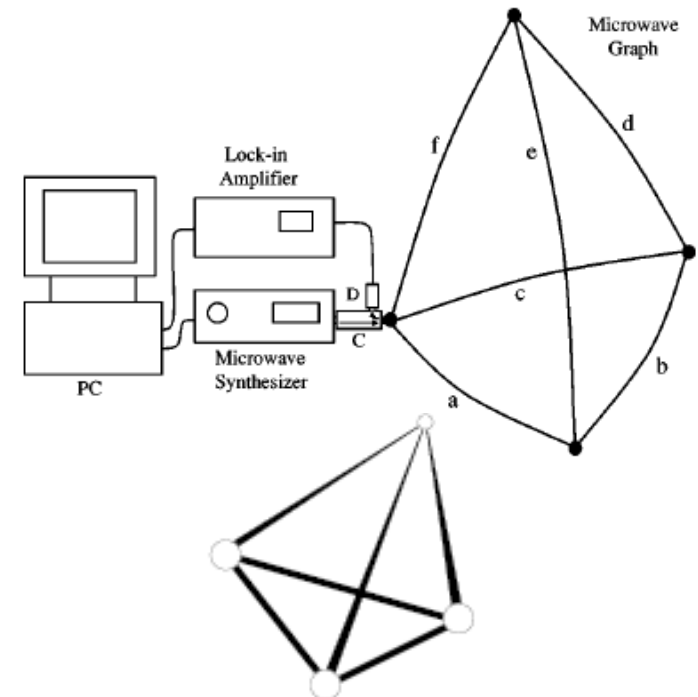
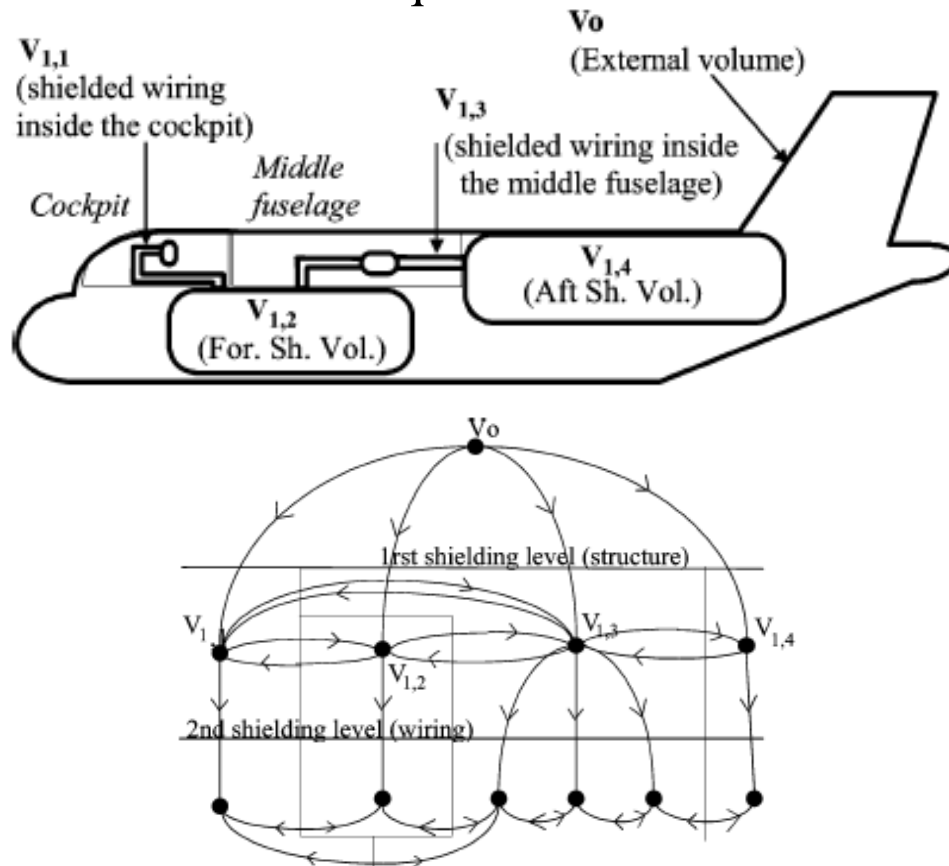
Bow-Tie with diamond scar

Ref: Antonsen et al., Phys. Rev E 51, 111 (1995).



Electromagnetic Topology and Wave Graphs

Electromagnetic Topology BLT Equations



J.-P. Parmantier, IEEE Trans. Electromag. Compat. 46 (3) 359-367 (2004). "Numerical coupling models for complex systems and results"

O. Hul, et al., Phys. Rev. E 69, 056205 (2004). "Experimental simulation of quantum graphs by microwave networks"